# Hybrid Precoding for Millimeter Wave MIMO Systems

# - Algorithm Design and Hardware Implementation





#### **Collaborators**



Xianghao Yu



Khaled B. Letaief

Hybrid Precoding for mm-Wave MIMO Systems Outline



# Background and Motivation

- Introduction to Hybrid Precoding
- Algorithm Design & Hardware Implementation
  - Single Phase Shifter (SPS) Implementation
  - Double Phase Shifter (DPS) Implementation
  - Fixed Phase Shifter (FPS) Implementation

# Conclusions

# Hybrid Precoding for mm-Wave MIMO Systems Background and Motivation



# Spectrum crunch: A fundamental bottleneck



mm-wave bands: Less congested, more bandwidth

# Hybrid Precoding for mm-Wave MIMO Systems Background and Motivation





[G. R. MacCartney et al., 2013]

Hybrid Precoding for mm-Wave MIMO Systems Background and Motivation





Large-scale antennas can be patched together

> Large antenna gain to compensate the path loss

Conventional Approach: Analog beamforming

State-of-art in mm-wave WiGig systems [E. Perahia et al., 2010]





New transceiver architecture

Sub-6 GHz systems: fully digital precoder

mm-wave systems: hybrid precoder





**QI**: Can it approach the performance of the fully digital precoding?

- Key differentiating component
- Q2: How many RF chains are needed?
- > Mapping from RF chains to antennas



Analog RF precoder structure

> Signal flow determines the "mapping strategy"



Q3: How to connect RF chains and antennas?



Analog RF precoder structure (cont'd)

- > Adopted hardware determines the "implementation"
- $\succ$  For each connected signal flow



**Q4**: How many phase shifters are needed?



## General multiuser multicarrier systems



> One single digital precoder  $\mathbf{F}_{BBk,f}$  for each user on each subcarrier > Analog precoder  $\mathbf{F}_{RF}$  is shared by all the users and subcarriers

# Problem formulation

Minimize the Euclidean distance between the hybrid precoders and the fully digital precoder [O. El Ayach et al., 2014]



 $\succ$   $\mathcal{A}_x$  varies according to different mappings and implementations

$$\mathbf{F}_{\text{opt}} = \left[\mathbf{F}_{\text{opt}_{1,1}}, \cdots, \mathbf{F}_{\text{opt}_{k,f}}, \cdots, \mathbf{F}_{\text{opt}_{K,F}}\right] \in N_{\text{t}} \times KN_{s}F$$
$$\mathbf{F}_{\text{BB}} = \left[\mathbf{F}_{\text{BB}_{1,1}}, \cdots, \mathbf{F}_{\text{BB}_{k,f}}, \cdots, \mathbf{F}_{\text{BB}_{K,F}}\right] \in N_{\text{RF}}^{\text{t}} \times KN_{s}F$$

**Q4**: How to efficiently design hybrid precoding algorithms?

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## Existing works

- Most focused on the SPS implementation
  - Orthogonal Matching Pursuit (OMP) [O. El Ayach et al., 2014] [T. E. Bogale et al., 2014]
    - ${\boldsymbol{\diamondsuit}}$  A candidate set for  ${\bf F}_{\rm RF}$
    - Array response vectors or DFT matrix
  - Channel phase extraction [L. Liang et al., 2014]
  - Successive interference cancellation (SIC) [X. Gao et al., 2016]
- How to achieve the fully digital precoder [E. Zhang et al., 2014] [T. E. Bogale et al., 2016]
  - Large numbers of RF chains and PSs needed



Performance metrics

"Scoring triangle"



Baseline: SPS fully-connected with OMP

Hybrid Precoding for mm-Wave MIMO Systems Key Questions



QI: Can hybrid precoder provide performance close to the fully digital one?

- Q2: How many RF chains are needed?
- Q3: How to connect the RF chains and antennas?
- Q4: How many phase shifters are needed?
- Q5: How to efficiently design hybrid precoding algorithms?

Hybrid Precoding for mm-Wave MIMO Systems Key Design Aspects



Hardware complexity (# hardware components)

Computational efficiency of precoding algorithms

Achievable spectrum efficiency



Single phase shifter (SPS) implementation



Start from single-user systems

Alternating minimization

- $N = \begin{cases} N_{\rm t} & \text{fully-connected} \\ N_{\rm t}/N_{\rm RF}^{\rm t} & \text{partially-connected} \end{cases}$
- Fully digital achieving condition:  $N_{\rm RF}^{\rm t} = 2KN_s, N_{\rm RF}^{\rm r} = 2N_s$

Q: Can we further reduce the number of RF chains?

[Ref] X. Yu, J.-C. Shen, J. Zhang, and K. B. Letaief, "Alternating minimization algorithms for hybrid precoding in millimeter wave MIMO systems," *IEEE J. Sel. Topics Signal Process.*, vol. 10, no. 3, pp. 485-500, Apr. 2016.



Fully-connected mapping

 $\underset{\mathbf{F}_{\mathrm{RF}}}{\operatorname{minimize}} \quad \left\| \mathbf{F}_{\mathrm{opt}} - \mathbf{F}_{\mathrm{RF}} \mathbf{F}_{\mathrm{BB}} \right\|_{F}^{2}$ 

> Digital precoder:  $\mathbf{F}_{BB} = \mathbf{F}_{RF}^{\dagger} \mathbf{F}_{opt}$ 

> Difficulty: Analog precoder design with the unit modulus constraints

 $\begin{array}{ll} \underset{\mathbf{F}_{\mathrm{RF}},\mathbf{F}_{\mathrm{BB}}}{\text{minimize}} & \|\mathbf{F}_{\mathrm{opt}} - \mathbf{F}_{\mathrm{RF}}\mathbf{F}_{\mathrm{BB}}\|_{F}^{2} \\ \text{subject to} & |(\mathbf{F}_{\mathrm{RF}})_{i,j}| = 1, \forall i, j. \end{array}$ 

The vector  $\mathbf{x} = \operatorname{vec}(\mathbf{F}_{\mathrm{RF}})$  forms a complex circle manifold  $\mathcal{M}^m = \{\mathbf{x} \in \mathbb{C}^m : |\mathbf{x}_1| = |\mathbf{x}_2| = \cdots = |\mathbf{x}_m| = 1\}, \quad m = N_{\mathrm{t}} N_{\mathrm{RF}}^{\mathrm{t}}.$ 

3 key elements in manifold optimization

> Tangent space:

$$T_{\mathbf{x}}\mathcal{M}^m = \{\mathbf{y} \in \mathbb{C}^m : \Re \{\mathbf{y} \circ \mathbf{x}^*\} = \mathbf{0}_m\}$$

Riemannian gradient

$$\begin{split} \mathrm{grad} f(\mathbf{x}) &= \mathrm{Proj}_{\mathbf{x}} \nabla f(\mathbf{x}) \\ &= \nabla f(\mathbf{x}) - \Re\{\mathrm{diag}\left[\nabla f(\mathbf{x}) \circ \mathbf{x}^*\right]\}\mathbf{x} \end{split}$$

$$\nabla f(\mathbf{x}) = -2(\mathbf{F}_{\mathrm{BB}}^* \otimes \mathbf{I}_{N_t}) \left[ \operatorname{vec}(\mathbf{F}_{\mathrm{opt}}) - (\mathbf{F}_{\mathrm{BB}}^T \otimes \mathbf{I}_{N_t}) \mathbf{x} \right]$$

- $\blacktriangleright \text{Retraction:} \quad \operatorname{Retr}_{\mathbf{x}} : T_{\mathbf{x}} \mathcal{M}^m \to \mathcal{M}^m :$  $\alpha \mathbf{d} \mapsto \operatorname{Retr}_{\mathbf{x}}(\alpha \mathbf{d}) = \operatorname{vec}\left[\frac{(\mathbf{x} + \alpha \mathbf{d})_i}{|(\mathbf{x} + \alpha \mathbf{d})_i|}\right]$
- Conjugate gradient algorithm
  - Local optimum guaranteed









## Partially-connected mapping

 $\succ$  Block diagonal structure of  $\mathbf{F}_{\mathrm{RF}}$ 

$$\mathbf{F}_{\mathrm{RF}} = \begin{bmatrix} \mathbf{p}_{1} & \mathbf{0} & \cdots & \mathbf{0} \\ \mathbf{0} & \mathbf{p}_{2} & \mathbf{0} \\ \vdots & \ddots & \vdots \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{p}_{N_{\mathrm{RF}}^{\mathrm{t}}} \end{bmatrix}$$
$$\mathbf{p}_{i} = \left[ \exp\left(j\theta_{(i-1)\frac{N_{t}}{N_{\mathrm{RF}}^{t}}} + 1\right), \cdots, \exp\left(j\theta_{i\frac{N_{t}}{N_{\mathrm{RF}}^{t}}}\right) \right]^{T}$$

phase siniter's connected to the i-th KF chain

Problem decoupled for each RF chain

 $\succ$  Closed-form solution for  $\mathbf{F}_{\mathrm{RF}}$ 

$$\arg\left\{(\mathbf{F}_{\mathrm{RF}})_{i,l}\right\} = \arg\left\{(\mathbf{F}_{\mathrm{opt}})_{i,:}(\mathbf{F}_{\mathrm{BB}})_{l,:}\right\}, \quad 1 \le i \le N_t, l = \left[i\frac{N_{\mathrm{RF}}^t}{N_t}\right]$$



Partially-connected mapping (cont'd)

 $\succ$  Optimization of  $\mathbf{F}_{BB}$ 

 $\begin{array}{ll} \underset{\mathbf{F}_{\mathrm{BB}}}{\operatorname{minimize}} & \left\|\mathbf{F}_{\mathrm{opt}} - \mathbf{F}_{\mathrm{RF}} \mathbf{F}_{\mathrm{BB}}\right\|_{F}^{2} \\ \text{subject to} & \left\|\mathbf{F}_{\mathrm{BB}}\right\|_{F}^{2} = \frac{N_{\mathrm{RF}}^{t} N_{s}}{N_{t}}. \end{array}$ 

Reformulate as a non-convex QCQP problem

$$\begin{array}{ll} \underset{\mathbf{Y} \in \mathbb{H}^{n}}{\text{minimize}} & \operatorname{Tr}(\mathbf{CY}) \\ \text{subject to} & \begin{cases} \operatorname{Tr}(\mathbf{A}_{1}\mathbf{Y}) = \frac{N_{\mathrm{RF}}^{t}N_{s}}{N_{t}} \\ \operatorname{Tr}(\mathbf{A}_{2}\mathbf{Y}) = 1 \\ \mathbf{Y} \succeq 0, \ \operatorname{rank}(\mathbf{Y}) = 1 \end{cases} \end{array}$$

SDR is tight so globally optimal solution is obtained [Z.-Q. Luo et al., 2010]

Converge to a local optimum

#### Simulation results





Effectiveness of the proposed AltMin algorithms

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The fully-connected mapping can easily approach the performance of the fully digital precoding

#### Hybrid Precoding for mm-Wave MIMO Systems UNIVERSITY OF SCIENCE Algorithm Design & Hardware Implementation I: SPS

## Simulation results



 $N_{\rm t} = 144, N_{\rm r} = 36, N_{\rm RF}^{\rm t} = N_{\rm RF}^{\rm r} = N_{\rm RF}, N_s = 2, \text{SNR} = 0 \,\mathrm{dB}$ 

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Limitation: computational efficiency not good, thus difficult to extend to MU multicarrier settings

Double phase shifter (DPS) implementation



 $N = \begin{cases} N_{\rm t} & \text{fully-connected} \\ N_{\rm t}/N_{\rm RF}^{\rm t} & \text{partially-connected} \end{cases}$ 

Q:What is the benefit?

Sum of two phase shifters  $|e^{j\theta_1} + e^{j\theta_2}| \le 2$ 

[Ref] X. Yu, J. Zhang, and K. B. Letaief, "Doubling phase shifters for efficient hybrid precoder design in millimeter-wave communication systems," submitted to *IEEE Trans. Wireless Commun.*, Mar. 2017.

# Fully-connected mapping

#### RF-only precoding

 $\begin{array}{ll} \underset{\mathbf{F}_{\mathrm{RF}}}{\text{minimize}} & \|\mathbf{F}_{\mathrm{opt}} - \mathbf{F}_{\mathrm{RF}} \mathbf{F}_{\mathrm{BB}}\|_{F}^{2} \\ \text{subject to} & |(\mathbf{F}_{\mathrm{RF}})_{i,j}| \leq 2 \end{array} \xrightarrow{\text{minimize}} & \frac{1}{2} \|\mathbf{A}\mathbf{x} - \mathbf{b}\|_{2}^{2} + 2\|\mathbf{x}\|_{1} \\ & \mathbf{LASSO} \end{array}$ 

 $\succ \text{Closed-form solution for semi-unitary codebooks } \mathbf{F}_{BB}\mathbf{F}_{BB}^{H} = \mathbf{I}_{N_{RF}^{t}}$  $\mathbf{F}_{RF}^{\star} = \mathbf{F}_{opt}\mathbf{F}_{BB}^{H} - \exp\left\{j\angle\left(\mathbf{F}_{opt}\mathbf{F}_{BB}^{H}\right)\right\} \circ \left(|\mathbf{F}_{opt}\mathbf{F}_{BB}^{H}| - 2\right)^{+}.$ 

> Hybrid precoding

#### Redundant

Fully-connected mapping (cont'd)

> Optimality in single-carrier systems

 $\mathbf{F}_{\text{opt}} = \mathbf{F}_{\text{RF}} \mathbf{F}_{\text{BB}}$  with  $N_{\text{RF}}^{\text{t}} = KN_s$  and  $N_{\text{RF}}^{\text{r}} = N_s$  when F = 1

Minimum number of RF chains

Reduced the number of RF chains by half required for achieving the fully digital precoding

Multi-carrier systems

$$\begin{array}{l} \underset{\mathbf{F}_{\mathrm{RF}},\mathbf{F}_{\mathrm{BB}}}{\operatorname{minimize}} \quad \left\|\mathbf{F}_{\mathrm{opt}} - \mathbf{F}_{\mathrm{RF}}\mathbf{F}_{\mathrm{BB}}\right\|_{F}^{2} \\ & \searrow \text{ Low-rank matrix approximation: SVD} \\ & \searrow \text{ Optimal solution} \end{array}$$

#### Hybrid Precoding for mm-Wave MIMO Systems UNIVERSITY OF SCIENCE AND TECHNOLOGY Algorithm Design & Hardware Implementation II: DPS



Block diagonal structure

$$\mathbf{F}_{ ext{RF}} = egin{bmatrix} \mathbf{p}_1 & \mathbf{0} & \cdots & \mathbf{0} \ \mathbf{0} & \mathbf{p}_2 & \mathbf{0} \ dots & \ddots & dots \ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{p}_{N_{ ext{RF}}^t} \end{bmatrix}$$

Not much performance gain obtained by the DFS implementation in the partially-connected mapping

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$$\mathbf{p}_{j} = \left[a_{(j-1)\frac{N_{\mathrm{t}}}{N_{\mathrm{RF}}^{\mathrm{t}}}+1}, \cdots, a_{j\frac{N_{\mathrm{t}}}{N_{\mathrm{RF}}^{\mathrm{t}}}}\right]^{T}$$

Decoupled for each RF chain

$$\mathcal{P}_{j}: \quad \underset{\{a_{i}\},\mathbf{x}_{j}}{\operatorname{minimize}} \sum_{i \in \mathcal{F}_{j}} \|\mathbf{y}_{i} - a_{i}\mathbf{x}_{j}\|_{2}^{2},$$
$$\mathcal{F}_{j} = \left\{ i \in \mathbb{Z} \left| (j-1) \frac{N_{t}}{N_{\mathrm{RF}}^{t}} + 1 \leq i \leq j \frac{N_{t}}{N_{\mathrm{RF}}^{t}} \right\}, \ \mathbf{y}_{i} = \mathbf{F}_{\mathrm{opt}}^{T}(i,:), \ \text{and} \ \mathbf{x}_{j} = \mathbf{F}_{\mathrm{BB}}^{T}(j,:)$$
$$\blacktriangleright \text{ Eigenvalue problem} \qquad \mathbf{x}_{j}^{\star} = \boldsymbol{\lambda}_{1} \left( \sum_{i \in \mathcal{F}_{j}} \mathbf{y}_{i} \mathbf{y}_{i}^{H} \right), \quad a_{i}^{\star} = \frac{\mathbf{x}_{j}^{H} \mathbf{y}_{i}}{||\mathbf{x}_{j}||_{2}^{2}}$$

Partially-connected mapping (cont'd)

Dynamic mapping

Adaptively separate all the antennas into  $N_{RF}$  groups

Problem formulation

 $\begin{array}{ll} \underset{\{\mathcal{D}_{j}\}_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}}}{\mathrm{maximize}} & \sum_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}} \lambda_{1} \left( \sum_{i \in \mathcal{D}_{j}} \mathbf{y}_{i} \mathbf{y}_{i}^{H} \right) & \max_{\{\mathcal{D}_{j}, \mathbf{x}_{j}\}_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}}} & \sum_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}} \frac{\mathbf{x}_{j}^{H} \left( \sum_{i \in \mathcal{D}_{j}} \mathbf{y}_{i} \mathbf{y}_{i}^{H} \right) \mathbf{x}_{j}}{\mathbf{x}_{j} \mathbf{x}_{j}^{H}} \\ \text{subject to} & \begin{cases} \bigcup_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}} \mathcal{D}_{j} = \{1, \cdots, N_{\mathrm{t}}\} \\ \mathcal{D}_{j} \cap \mathcal{D}_{k} = \emptyset, \quad \forall j \neq k \end{cases} & \text{subject to} & \begin{cases} \bigcup_{j=1}^{N_{\mathrm{RF}}^{\mathrm{t}}} \mathcal{D}_{j} = \{1, \cdots, N_{\mathrm{t}}\} \\ \mathcal{D}_{j} \cap \mathcal{D}_{k} = \emptyset, \quad \forall j \neq k \end{cases} \end{array}$ 

Modified K-means algorithm

Centroid: 
$$\mathbf{x}_j^{\star} = \boldsymbol{\lambda}_1 \left( \sum_{i \in \mathcal{D}_j} \mathbf{y}_i \mathbf{y}_i^H \right)$$

Converge to a local optimum

Clustering: 
$$j^{\star} = \arg \max_{j} |\mathbf{y}_{i}^{H}\mathbf{x}_{j}|^{2}$$

#### Inter-user interference

- Approximating the fully digital precoder leads to a near-optimal performance in single-user single-carrier, single-user multicarrier, and multiuser single-carrier mmWave MIMO systems.
- Inter-user interference will be more prominent in the multicarrier system as the analog precoder is shared by a large number of subcarriers
- ➤ Cascade an additional block diagonalization (BD) precoder
   Effective channel: ÎH<sub>k,f</sub> = W<sup>H</sup><sub>BBk,f</sub>W<sup>H</sup><sub>RFk</sub>H<sub>k,f</sub>F<sub>RF</sub>F<sub>BBf</sub>
   BD: ÎH<sub>j,f</sub>F<sub>BDk,f</sub> = 0, k ≠ j

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spectral efficiency and optimal multiplexing gain with low-complexity

Effectiveness of the proposed DPS implementation and cascaded BD precoder

#### Simulation results (Partially-connected)

 $N_{\rm t} = 256, N_{\rm r} = 16, K = 4, F = 128, N_s = 2, N_{\rm RF}^{\rm t} = 8, \text{ and } N_{\rm RF}^{\rm r} = 2$ 



Simply doubling PSs in the partially-connected mapping is far from satisfactory

Superiority of the modified K-means algorithm with lower complexity than the greedy one





Fixed phase shifter (FPS) implementation



[Ref] X. Yu, J. Zhang, and K. B. Letaief, "Hybrid precoding in millimeter wave systems: How many phase shifters are needed?" accepted to *IEEE Global Commun. Conf. (GLOBECOM), Singapore, Dec. 2017.* 

# Problem formulation

$$\begin{array}{l} \underset{\mathbf{S},\mathbf{F}_{\mathrm{BB}}}{\text{minimize}} & \|\mathbf{F}_{\mathrm{opt}} - \mathbf{SCF}_{\mathrm{BB}}\|_{F}^{2} \\ \text{subject to} & \mathbf{S} \in \mathcal{B} \end{array} \\ \\ & \searrow \text{ FPS matrix } \mathbf{C} = \operatorname{diag}\left(\underbrace{\mathbf{c},\mathbf{c},\cdots,\mathbf{c}}_{N_{\mathrm{RF}}^{\mathrm{t}}}\right), \quad \mathbf{c} = \frac{1}{\sqrt{N_{c}}} \left[e^{j\theta_{1}},e^{j\theta_{2}},\cdots,e^{j\theta_{N_{c}}}\right]^{T} \end{array}$$

 $\blacktriangleright$  Switch matrix  $\mathbf{S} \in \{0, 1\}^{N_{\mathrm{t}} \times N_c N_{\mathrm{RF}}^{\mathrm{t}}}$ 

- An objective upper bound enables a low-complexity algorithm
  - $\succ \text{ Enforce a semi-orthogonal constraint of } \mathbf{F}_{\text{BB}} \text{ [X.Yu et al., 2016]}$  $\mathbf{F}_{\text{BB}}^{H} \mathbf{F}_{\text{BB}} = \alpha^{2} \mathbf{F}_{\text{DD}}^{H} \mathbf{F}_{\text{DD}} = \alpha^{2} \mathbf{I}_{KN_{s}}$  $\|\mathbf{F}_{\text{opt}} \mathbf{SCF}_{\text{BB}}\|_{F}^{2} \leq \|\mathbf{F}_{\text{opt}}\|_{F}^{2} 2\alpha \Re \operatorname{Tr} \left(\mathbf{F}_{\text{DD}} \mathbf{F}_{\text{opt}}^{H} \mathbf{SC}\right) + \alpha^{2} \|\mathbf{S}\|_{F}^{2}$

Hybrid Precoding for mm-Wave MIMO Systems UNIVERSITY OF SCIENCE AND TECHNOLOGY Algorithm Design & Hardware Implementation III: FPS



Digital precoder

 $\underset{\mathbf{F}_{\mathrm{DD}}}{\operatorname{maximize}} \quad \Re \operatorname{Tr} \left( \mathbf{F}_{\mathrm{DD}} \mathbf{F}_{\mathrm{opt}}^{H} \mathbf{SC} \right)$ subject to  $\mathbf{F}_{\text{DD}}^{H}\mathbf{F}_{\text{DD}} = \mathbf{I}_{KN_s}$ 

 $\succ$  Semi-orthogonal Procrustes solution  $\mathbf{F}_{DD} = \mathbf{V}_1 \mathbf{U}^H$ 

Switch matrix optimization

 $\underset{\alpha,\mathbf{S}}{\text{minimize}} \quad \left\| \Re \left( \mathbf{F}_{\text{opt}} \mathbf{F}_{\text{DD}}^{H} \mathbf{C}^{H} \right) - \alpha \mathbf{S} \right\|_{F}^{2}$ subject to  $\mathbf{S} \in \mathcal{B}$ 

 $\succ$  Once  $\alpha$  is optimized, the optimal **S** is determined correspondingly

$$\mathbf{S}^{\star} = \begin{cases} \mathcal{I} \left\{ \Re \left( \mathbf{F}_{\text{opt}} \mathbf{F}_{\text{DD}}^{H} \mathbf{C}^{H} \right) > \frac{\alpha}{2} \mathbf{1}_{N_{\text{t}} \times N_{c} N_{\text{RF}}^{\text{t}}} \right\} & \alpha > 0 \\ \mathcal{I} \left\{ \Re \left( \mathbf{F}_{\text{opt}} \mathbf{F}_{\text{DD}}^{H} \mathbf{C}^{H} \right) < \frac{\alpha}{2} \mathbf{1}_{N_{\text{t}} \times N_{c} N_{\text{RF}}^{\text{t}}} \right\} & \alpha < 0 \end{cases}$$

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Alternating minimization (cont'd)

 $\succ$  Optimization of  $\alpha$ 

$$\alpha^{\star} = \arg \min_{\{\tilde{x}_i, \bar{x}_i\}_{i=1}^n} \{f(\tilde{x}_i), f(\bar{x}_i)\}$$
$$\tilde{\mathbf{x}} = \operatorname{sort}(\operatorname{vec}(\mathbf{F}_{\operatorname{opt}} \mathbf{F}_{\operatorname{DD}}^H \mathbf{C}^H)) \quad \bar{x}_i \triangleq \begin{cases} \frac{\sum_{j=1}^i \tilde{x}_j}{n-i} & \alpha < 0 \text{ and } \frac{\sum_{j=1}^i \tilde{x}_j}{n-i} \in [2\tilde{x}_i, 2\tilde{x}_{i+1}] \\ \frac{\sum_{j=i+1}^n \tilde{x}_j}{n-i} & \alpha > 0 \text{ and } \frac{\sum_{j=i+1}^n \tilde{x}_j}{n-i} \in [2\tilde{x}_i, 2\tilde{x}_{i+1}] \\ +\infty & \text{otherwiese} \end{cases}$$

 $\succ$  Search dimension:  $2N_{\rm t}N_{\rm RF}^{\rm t}N_c$ 

 $\succ$  Optimal point can only be finite  $\bar{x}_i$ 

$$\alpha^{\star} = \arg\min_{\bar{x}_i \in \mathcal{X}} \quad f(\bar{x}_i)$$

 $\succ$  Search dimension:  $|\mathcal{X}| \ll 2N_{\rm t}N_{\rm RF}^{\rm t}N_c$ 

Converge to a local optimum

Q: How to reduce # phase shifters?

- New mapping strategy
  - Group-connected mapping



$$\mathbf{F}_{\mathrm{RF}} = egin{bmatrix} \mathbf{R}_1 & \mathbf{0} & \cdots & \mathbf{0} \ \mathbf{0} & \mathbf{R}_2 & & \mathbf{0} \ dots & & \ddots & dots \ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{R}_\eta \end{bmatrix}$$

>  $\eta = I$ : fully-connected mapping >  $\eta = N_{RF}$ : partially-connected

 $\begin{array}{ll} \underset{\mathbf{F}_{i},\mathbf{B}_{i}}{\text{minimize}} & \|\mathbf{F}_{i}-\mathbf{R}_{i}\mathbf{B}_{i}\|_{F}^{2} \\ \text{subject to} & \mathbf{R}_{i} \in \mathcal{A}_{i}, \end{array}$ 

Designed by directly migrating the design for the fully-connected mapping





 $N_{\rm t} = 144, N_{\rm r} = 16, N_{\rm RF}^{\rm t} = N_{\rm RF}^{\rm r} = N_s = 4, \text{ and } N_c = 30$ 



37

#### Simulation results: MU-MC systems

 $N_{\rm t} = 144, N_{\rm r} = 16, K = 4, F = 128, N_s = 2, N_{\rm RF}^{\rm t} = 8, \text{ and } N_{\rm RF}^{\rm r} = 2$ 





#### Simulation results: How many PSs are needed?

 $N_{\rm t} = 256, N_{\rm r} = 16, K = 4, F = 128, N_s = 2, N_{\rm RF}^{\rm t} = 8, \text{ and } N_{\rm RF}^{\rm r} = 2$ 



#### Simulation results

 $N_{\rm t} = 256, N_{\rm r} = 16, K = 4, F = 128, N_s = 2, N_{\rm RF}^{\rm t} = 8, \text{ and } N_{\rm RF}^{\rm r} = 2$ 





Conclusion



Hybrid Precoding for mm-Wave MIMO Systems Takeaways



Q2: How many RF chains are needed? KN<sub>s</sub>

- Q3: How to connect the RF chains and antennas? Group-connected
- Q4: How many phase shifters are needed? ~I5 FPSs
- Q5: How to efficiently design hybrid precoding algorithms?
   Alternating minimization, implementation dependent

# Hybrid Precoding for mm-Wave MIMO Systems Takeaways



# Comparisons

Imple-	Structure	Design approach	Hardware complexity	Computational	Performance
mentation			(No. of phase shifters)	complexity	
SPS	Fully-connected	MO-AltMin	$N_{\rm RF}^{\rm t} N_{\rm t}$	Extremely high	$\checkmark \checkmark \checkmark$
	Partially-connected	SDR-AltMin	$N_{ m t}$	High	$\checkmark$
DPS	Fully-connected	Matrix decomposition	$2N_{\rm RF}^{\rm t}(N_{\rm t}-N_{\rm RF}^{\rm t})$	$\mathcal{O}\left({N_{\mathrm{RF}}^{\mathrm{t}}}^2 N_{\mathrm{t}}F\right)$	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$
	Partially-connected	Modified K-means	$2N_{\rm t}$	$\mathcal{O}\left(N{N_{\mathrm{RF}}^{\mathrm{t}}}^2 N_{\mathrm{t}}F\right)$	$\checkmark$
FPS		FPS-Alt-Min	$N_c \ll N_{\rm t}$	Medium	$\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{\sqrt{$

- The current SPS implementation is not a good choice due to the high design complexity caused by the strict hardware limitation
- The DPS implementation is an excellent candidate if highperformance hybrid precoders are required with low design complexity
- The FPS implementation finds a trade-off between the hardware and computational complexity, while with satisfactory performance





- ➢ Joint design with CSI acquisition
- Performance evaluation of hybrid precoding algorithms
- Further reduction in computational complexity
- Hardware implementation and testing
- Hybrid precoding with low-precision ADCs
- ▶ .....

# Hybrid Precoding for mm-Wave MIMO Systems References



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- X. Yu, **J. Zhang**, and K. B. Letaief, "Partially-connected hybrid precoding in mm-wave systems with dynamic phase shifter networks," in *SPAWC*, Hokkaido, Japan, Jul. 2017.
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